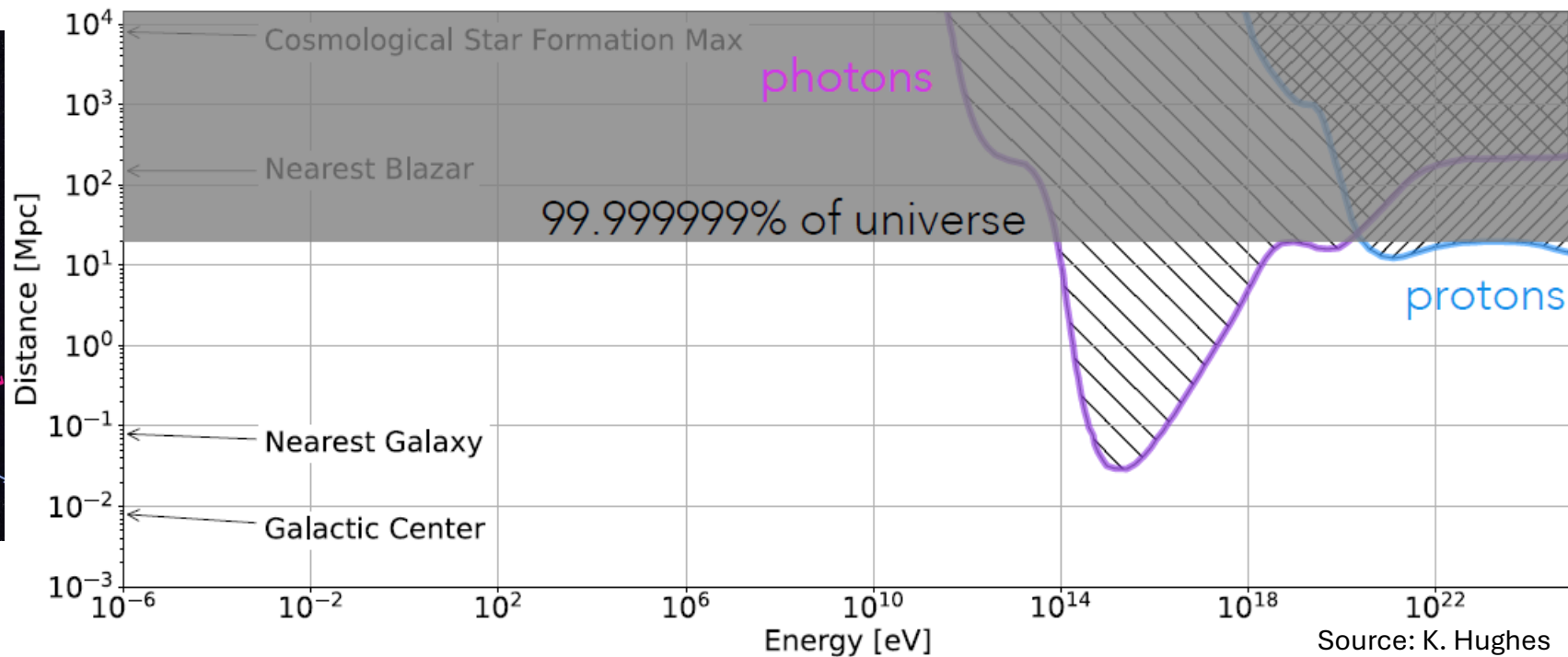
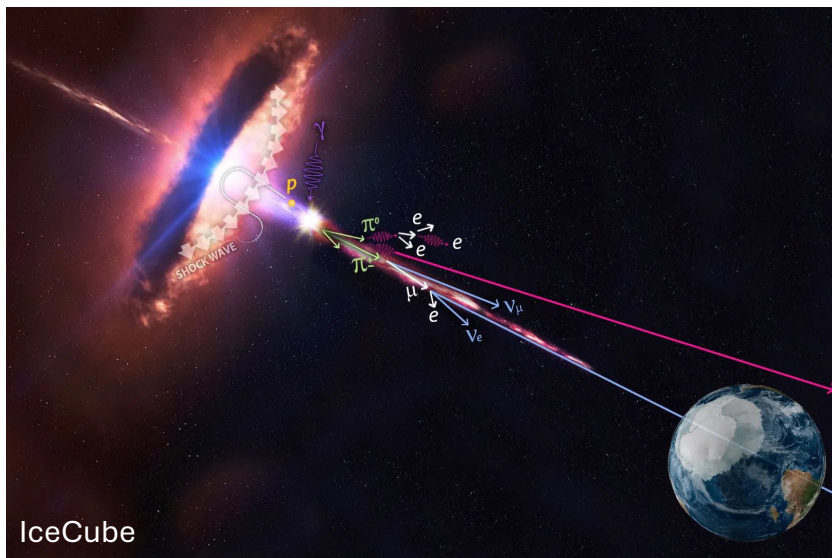


# Building the Payload for Ultrahigh Energy Observations

Lucas Beaufore  
The Ohio State University

We want to study the distribution and nature of the sources of the highest energy particles in the universe!

Neutrinos are uniquely well-suited messengers for this purpose.

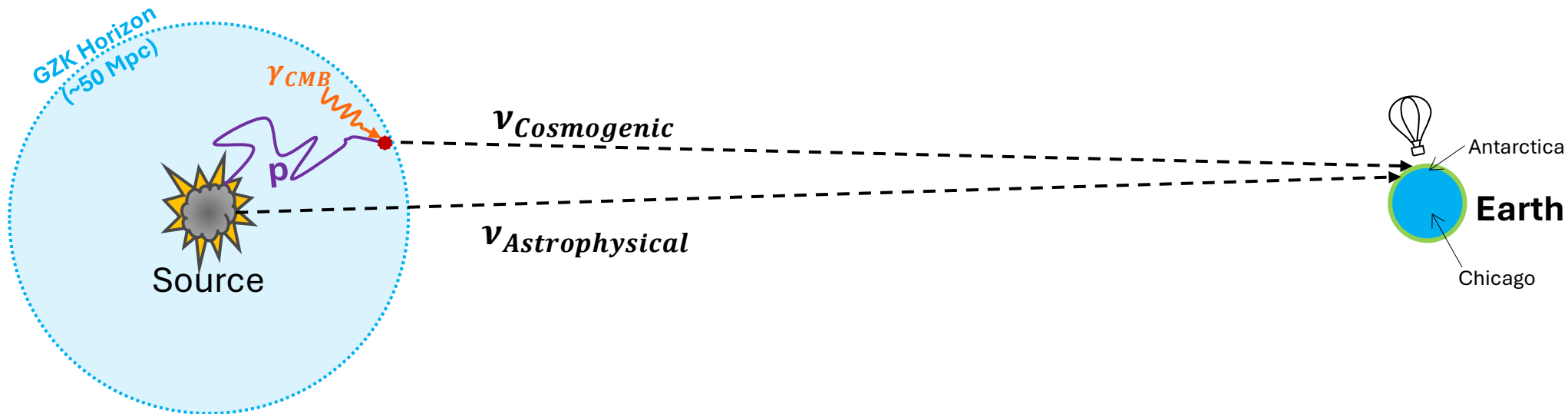


## Astrophysical

UHE neutrinos produced directly by their astrophysical sources.

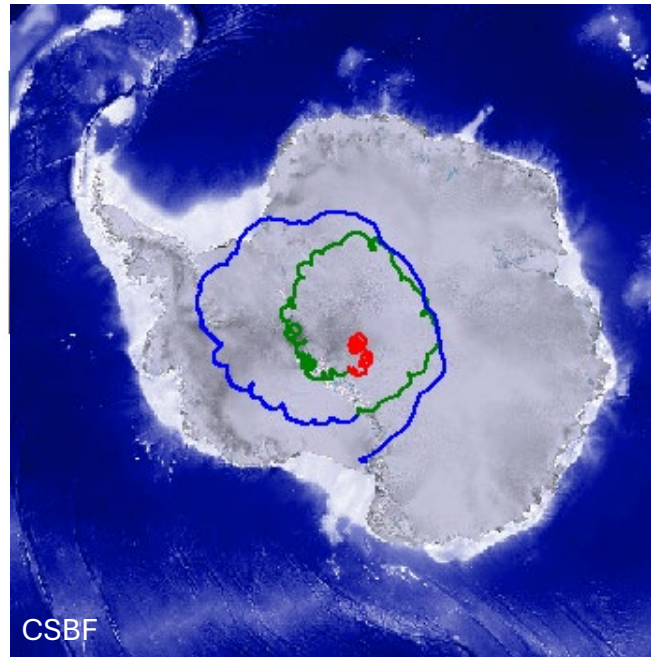
## Cosmogenic

UHECR with energies above  $\sim 50$  EeV interact with the CMB, producing neutrinos.

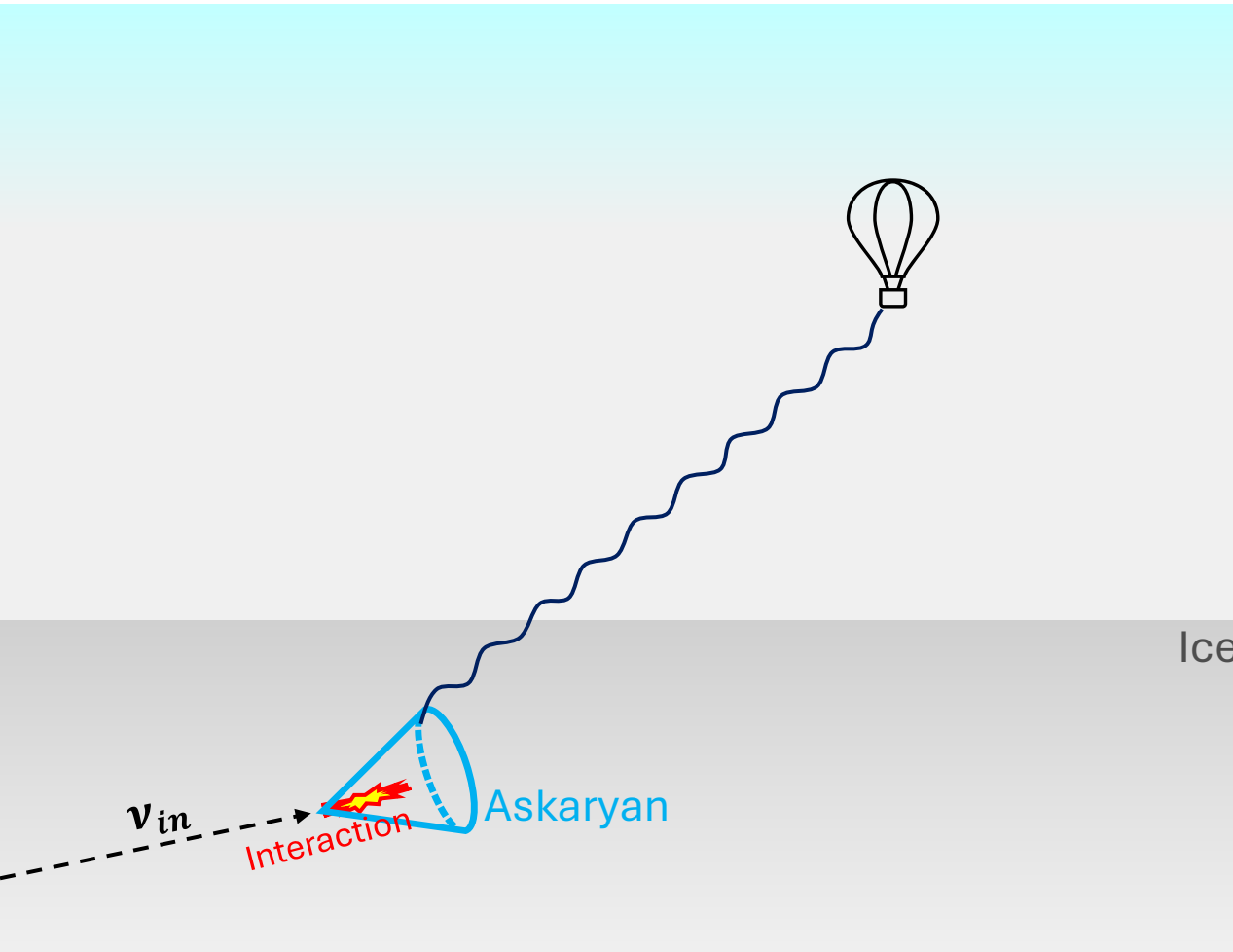


## Payload for **U**ltrahigh **E**nergy **O**bservations

- Radio detection experiment
- NASA Long Duration Balloon flight over Antarctica
- Will measure the ultrahigh energy neutrinos' interactions with the Earth
  - $>1 \text{ EeV}$  ( $10^{18} \text{ eV}$ )!



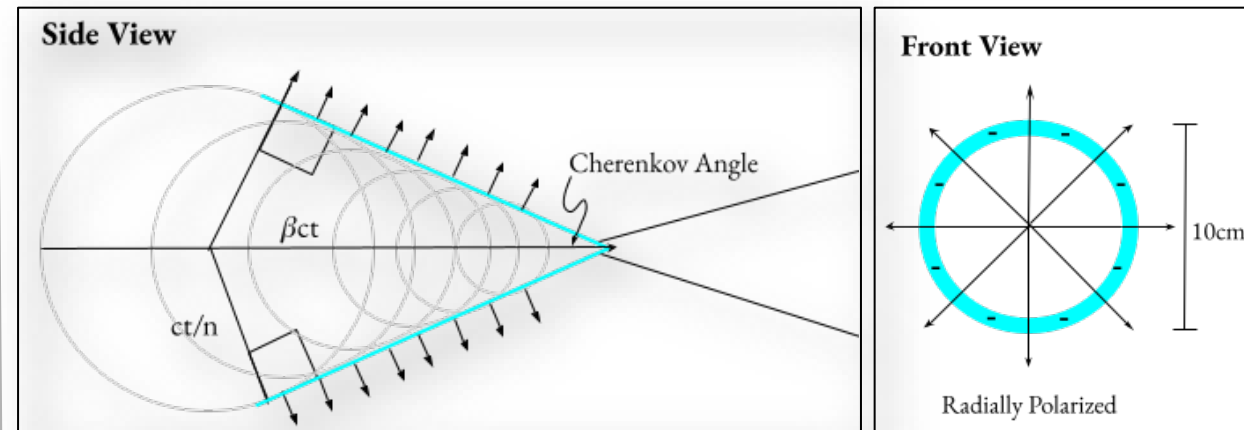
## Method 1: Askaryan emission in ice



When an UHE neutrino interacts in the ice, the resulting shower emits Askaryan radiation at the Cherenkov angle.

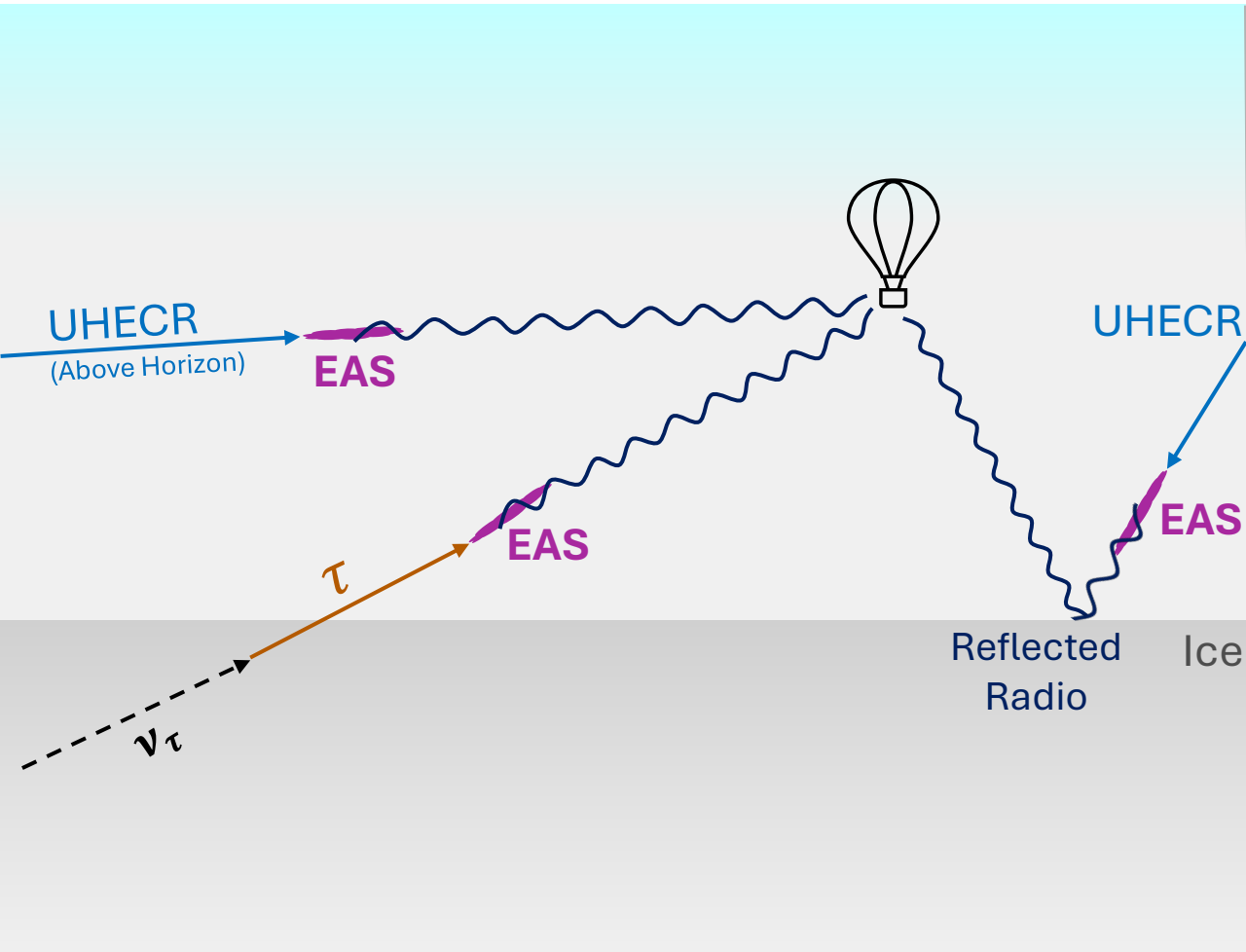
This radiation will add coherently in radio (with power  $\sim E^2$ ) if the width of the shower is less than radio wavelength.

The radio emission is radially polarized, which is useful for reconstructing the direction of the shower.



Credit: Rachel Scrandis

## Method 2: Geomagnetic emission in air showers

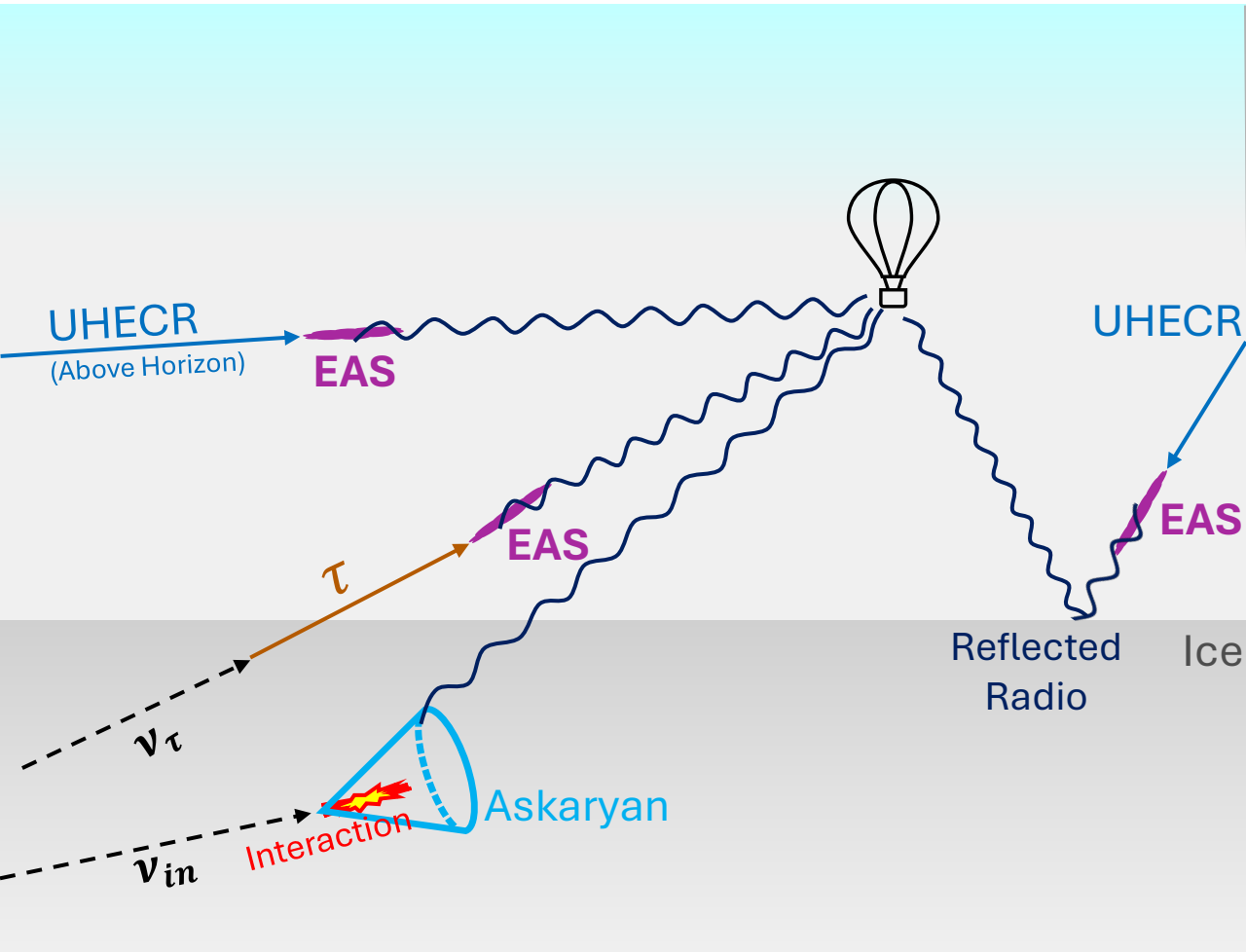


PUEO can also detect the radio that is produced by geomagnetic emission in air showers.

Tau neutrinos interacting in the earth produce tau particles, which then decay and produce an Extensive Air Shower (EAS).

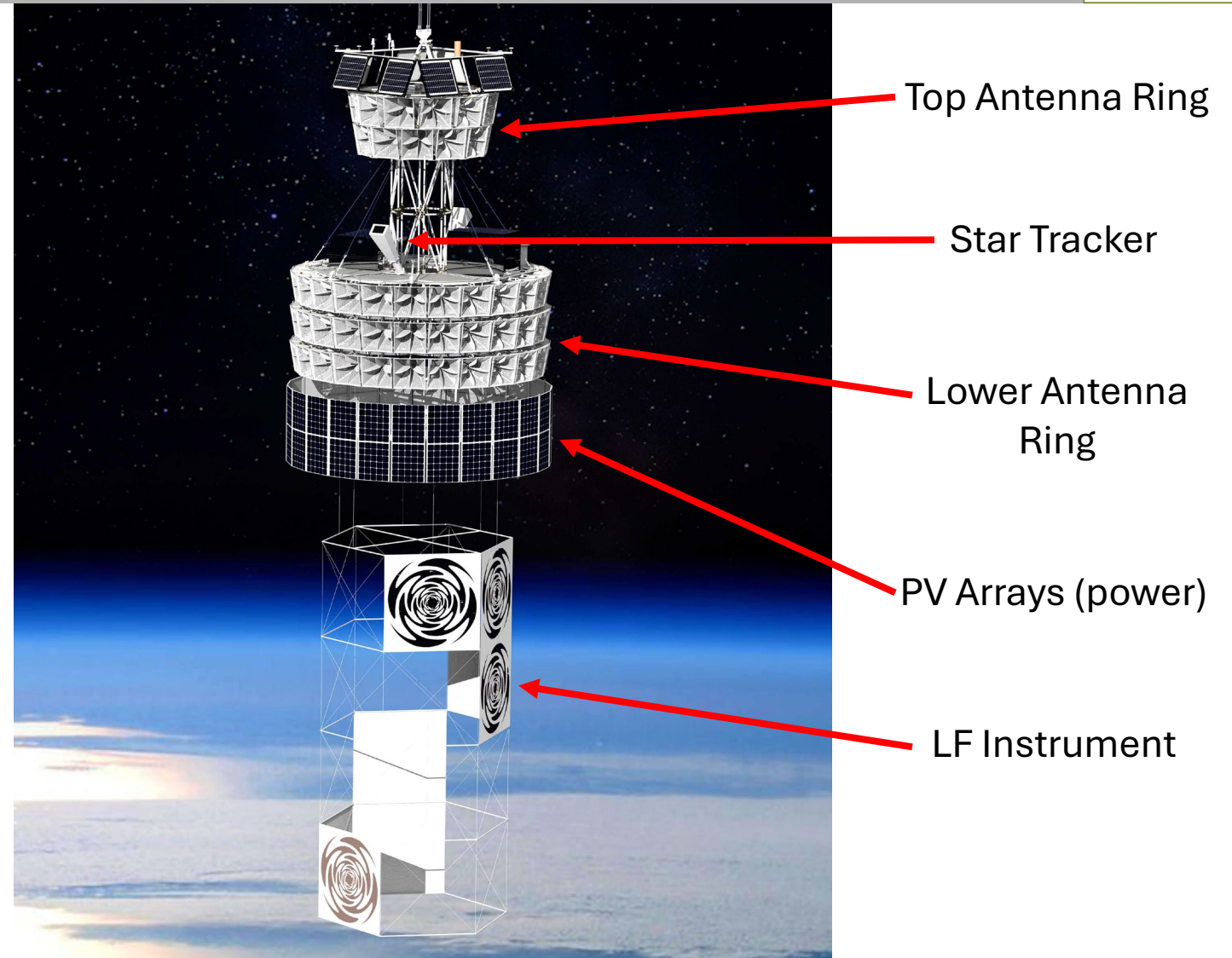
EASs can also be produced by UHE cosmic rays.

## Differentiation Between Radio Sources



- **In-ice Askaryan**
  - Below horizon
  - Vertically polarized
- **UHECR induced EAS**
  - Above horizon
  - Horizontally polarized
- **Tau-neutrino induced EAS**
  - Below horizon
  - Horizontally polarized
- **UHECR induced EAS (reflected)**
  - Below Horizon
  - Horizontally polarized
  - *Polarity flip relative to other EAS showers*

- Main instrument antennas
  - 96 Quad-ridged horns
  - Dual polarized
  - Arranged into 4 rings
  - $2\pi$  (full) azimuthal coverage
  - 300-1200 MHz band
- Low frequency instrument
  - 8 Sinuous antennas
  - Dual polarized
  - 50-500 MHz band
  - Made of conductive fabric



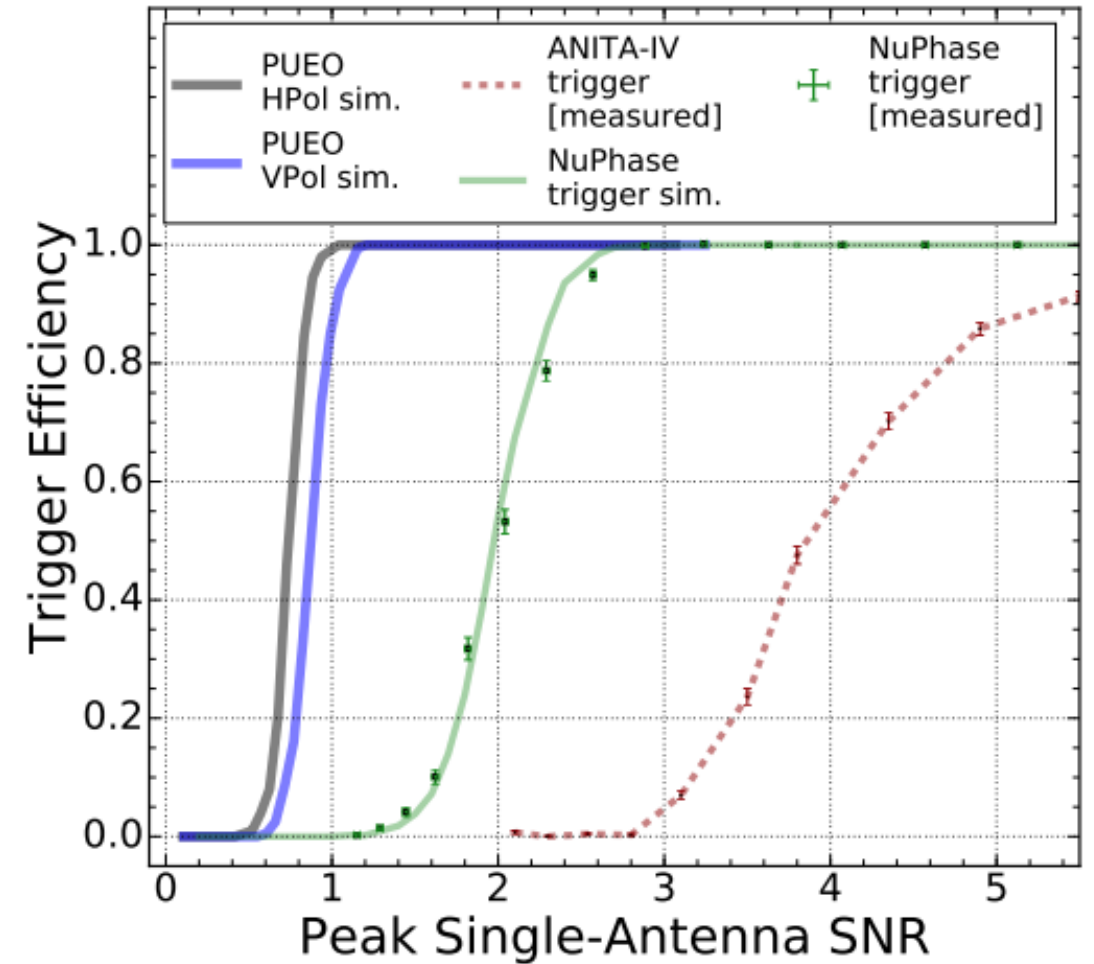




- 300 MHz high-pass antennas vs 200 MHz
  - Anthropogenic noise below 300 MHz
  - Smaller antennas → more collecting area
- Longer baselines between antennas
  - Improved pointing and background rejection
- Improved DAQ and trigger system
  - Phased array trigger → higher effective SNR
  - Enabled by use of RFSocCs (8 channels @ 3 GHz)
- LF instrument
  - Better sensitivity to air showers from CRs or taus
- Improved navigation/orientation suite

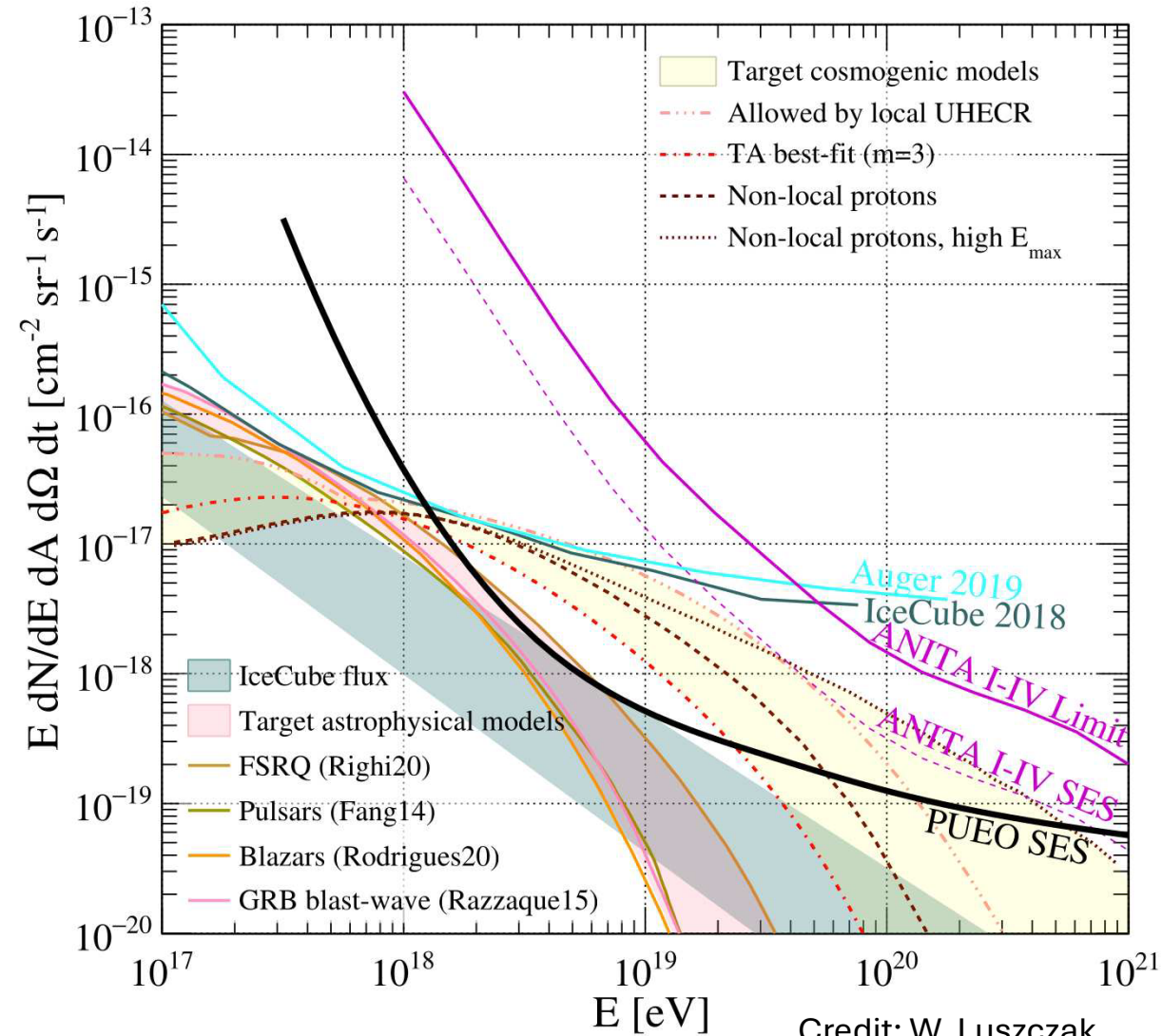


- RFSocS digitize signals from the antennas to 12 bits
- Real-time digital processing
  - Low-pass filter at  $\sim 750$  MHz
  - Tunable dual-biquad notch filters
  - Automatic gain control
  - 12-to-5 bits, then beamform
- Trigger thresholds adjust to maintain a constant event rate



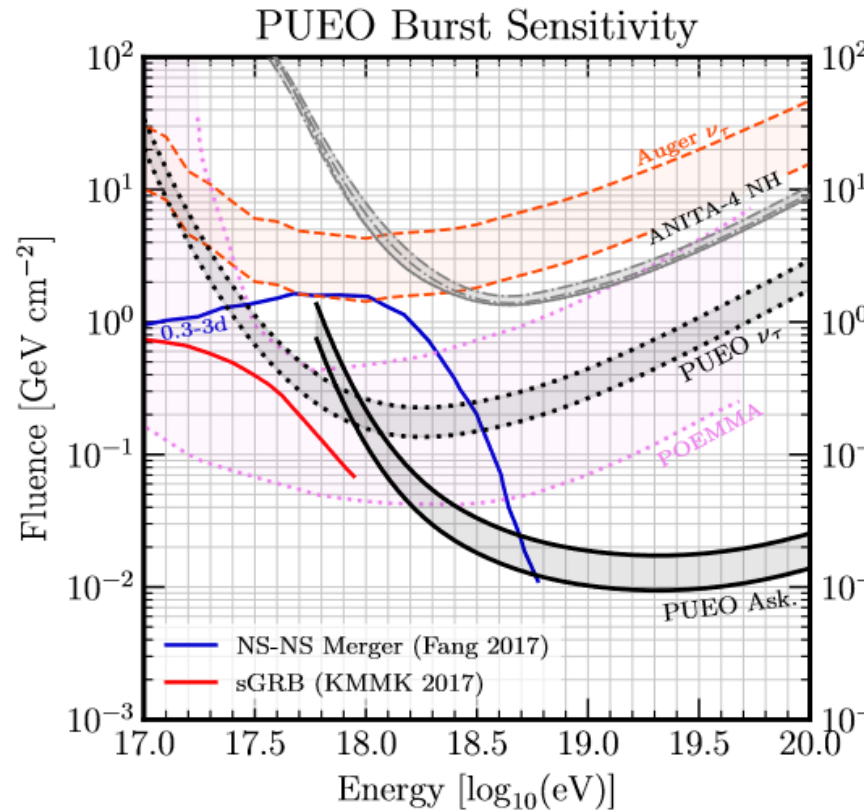
Pueo Whitepaper

- PUEO's single-event sensitivity (SES) to diffuse UHE fluxes will outperform the combined ANITA flights
  - Exclude or measure a number of cosmogenic models
- Multiple or longer flights could probe additional phase space
  - Astrophysical production models

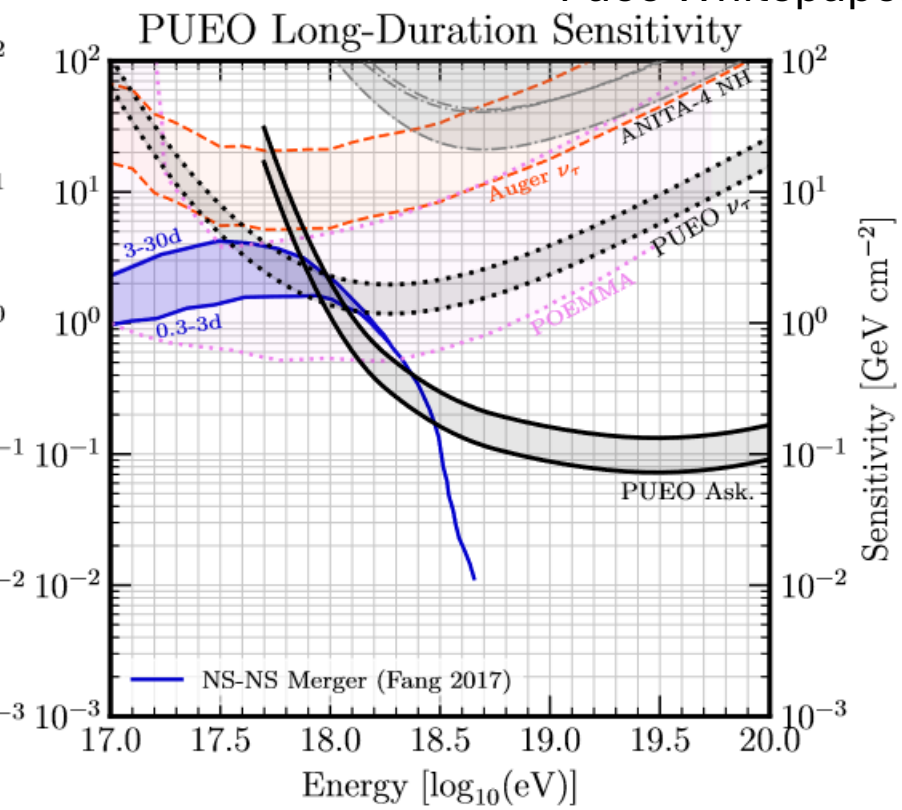


Credit: W. Luszczyk, ICRC 2023

PUEO's large instantaneous aperture makes it ideal for transient searches within its field-of view!



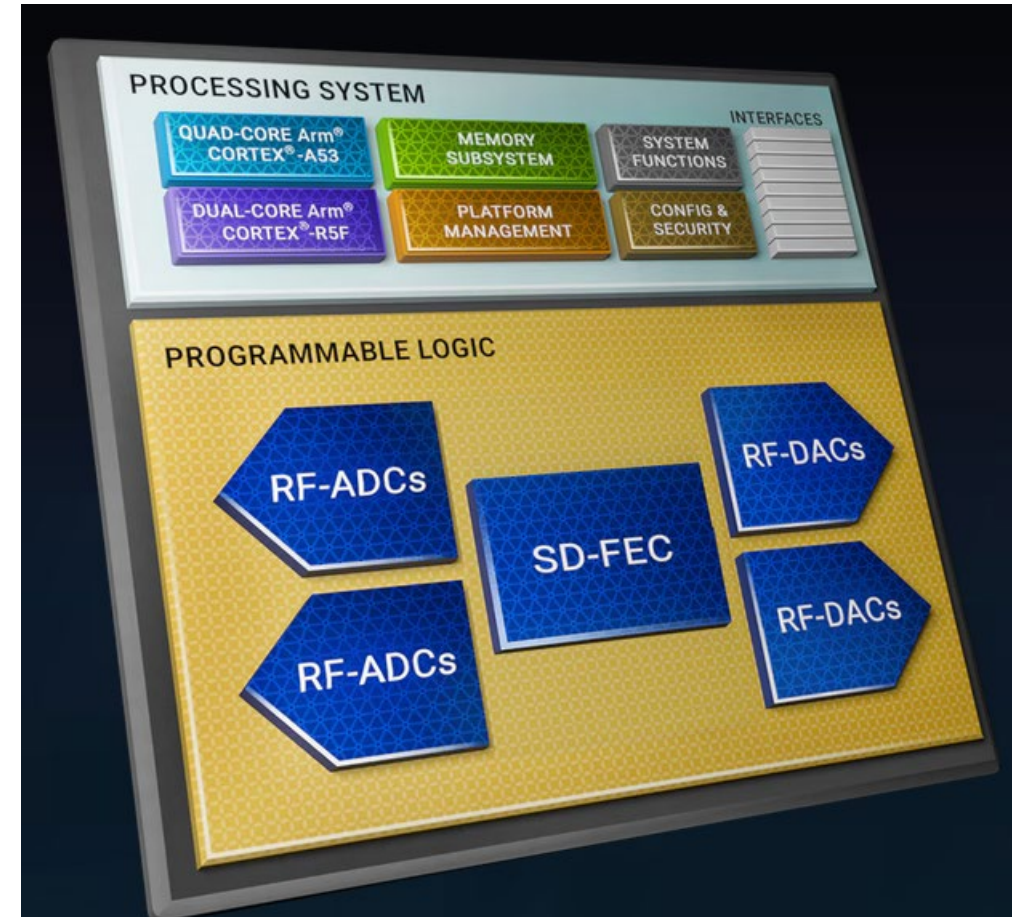
Burst sensitivity is for transients lasting a few hours.



Long duration sensitivity is for transients lasting around the same length as the flight, so effective area is averaged over the full flight.

Wait, Lucas, what do you actually  
do all day?

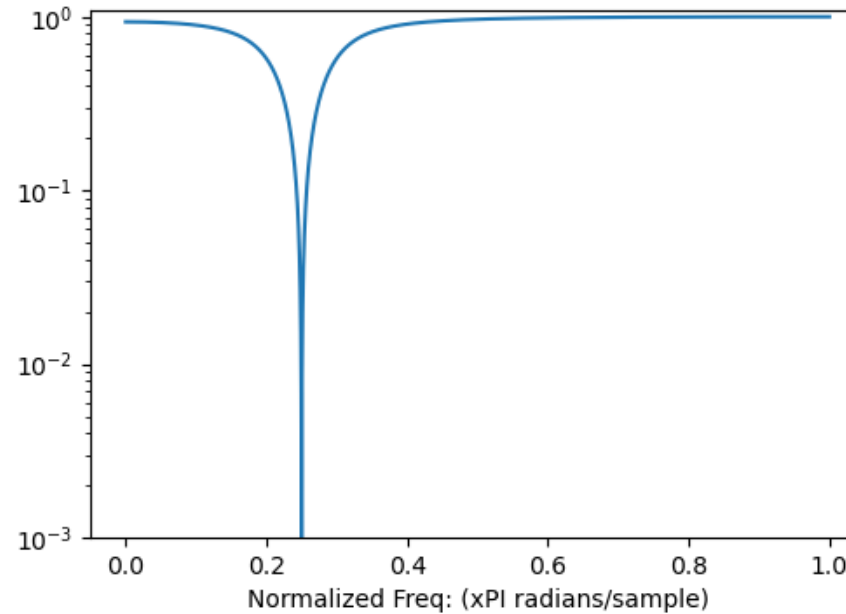
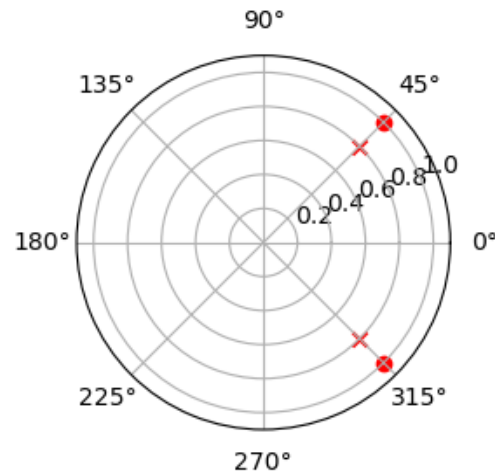
- Firmware, in this context, means FPGAs.
- Why use FPGAs (Field Programmable Gate Arrays)?
  - A serial processor, like the one in your computer, is not ideal for digitizing, filtering, triggering, and buffering a large and fast system!
  - A highly parallel digital circuit is much better suited for these applications.
  - FPGAs can implement these, and they are reprogrammable.



$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$

Zeros

Poles

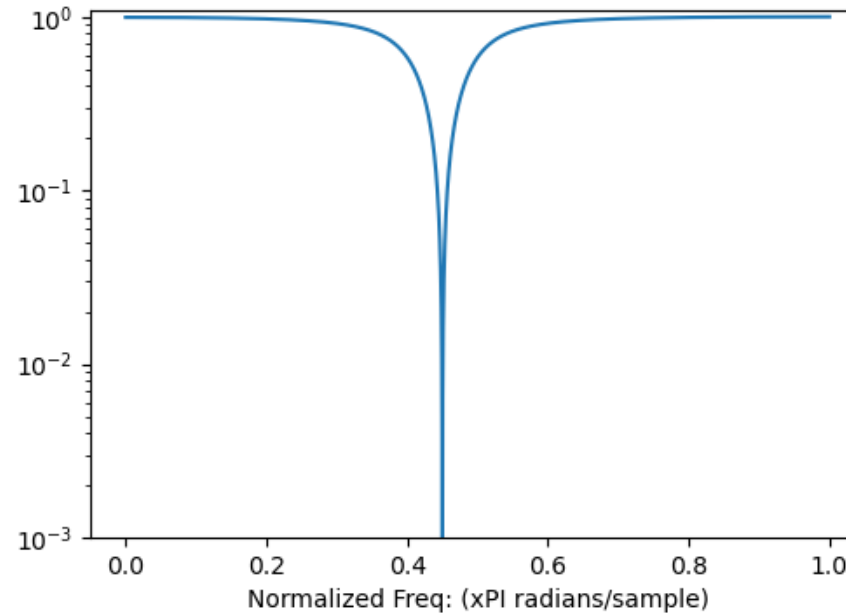
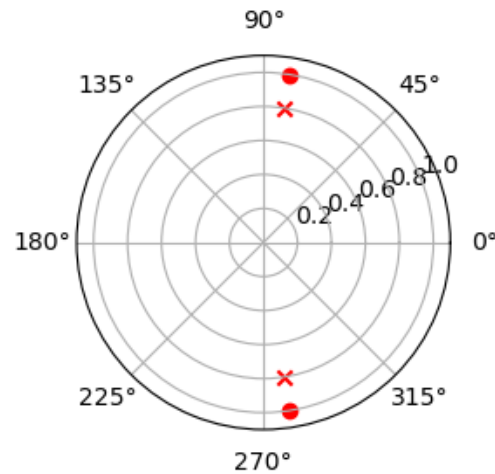


Plotting script adapted from Jim Beatty's in pueo-utilities

$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$

Zeros

Poles



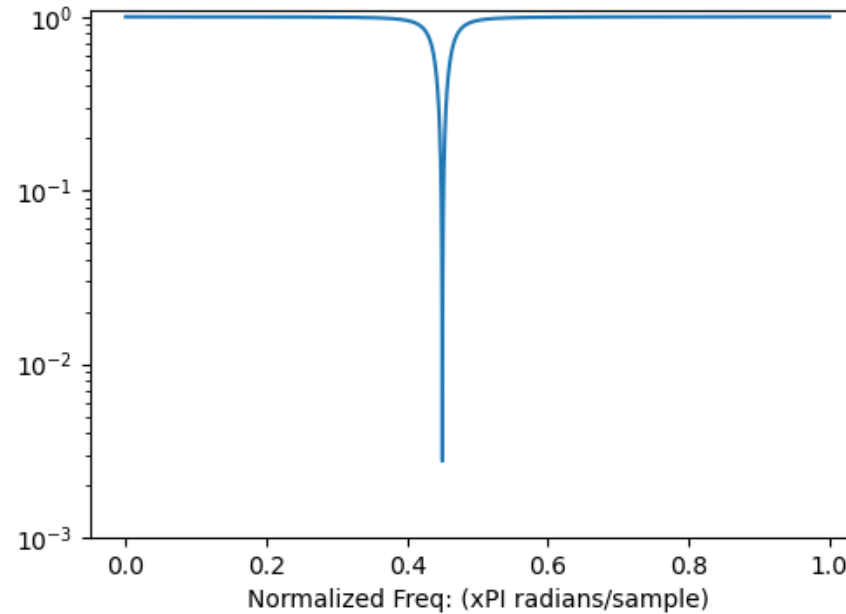
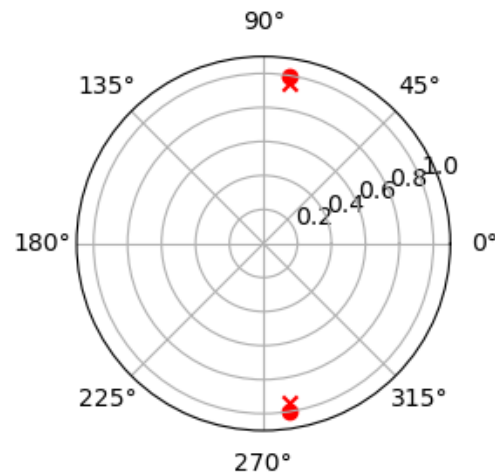
Plotting script adapted from Jim Beatty's in pueo-utilities



$$H(z) = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2}}{1 + a_1 z^{-1} + a_2 z^{-2}}$$

Zeros

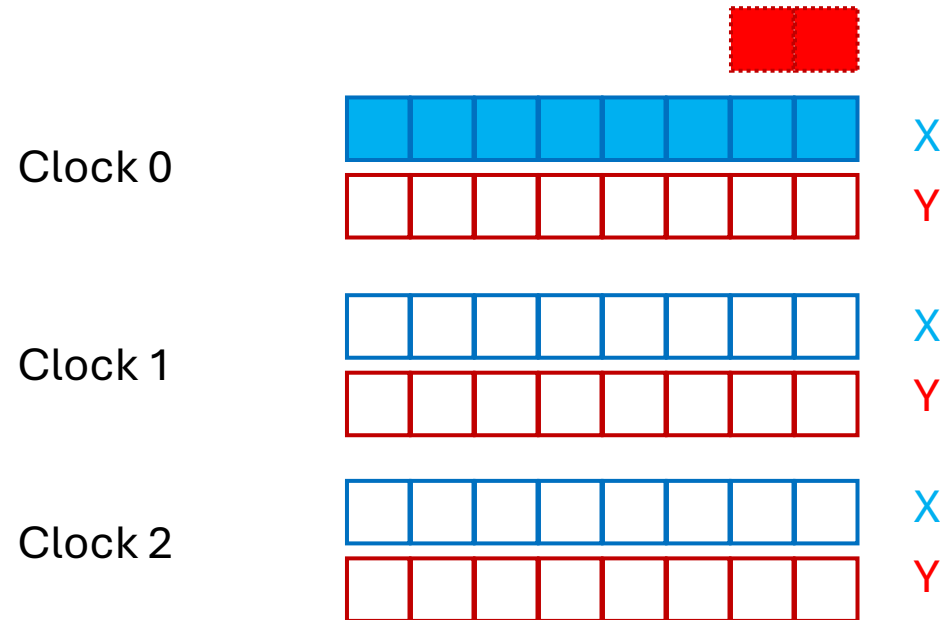
Poles



Plotting script adapted from Jim Beatty's in pueo-utilities

$$H(z) = \frac{b_0 + b_1z^{-1} + b_2z^{-2}}{1 + a_1z^{-1} + a_2z^{-2}}$$

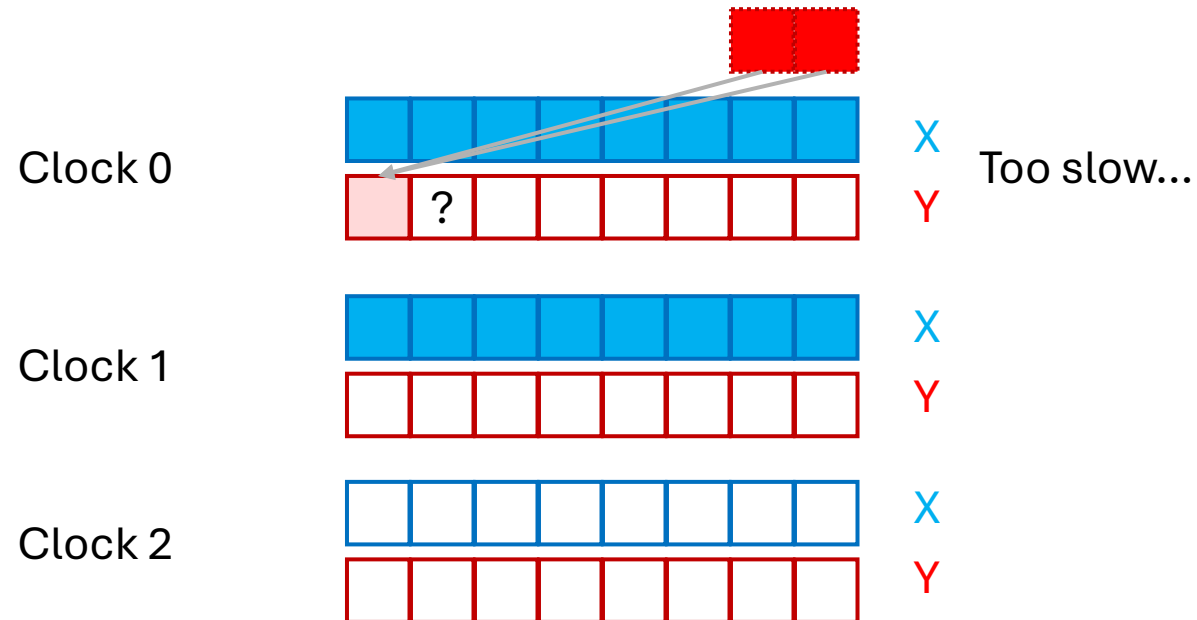
$$y[n] = b_0x[n] + b_1x[n - 1] + b_2x[n - 2] + a_1y[n - 1] + a_2y[n - 2]$$



\* Some latency ignored

$$H(z) = \frac{b_0 + b_1z^{-1} + b_2z^{-2}}{1 + a_1z^{-1} + a_2z^{-2}}$$

$$y[n] = b_0x[n] + b_1x[n - 1] + b_2x[n - 2] + a_1y[n - 1] + a_2y[n - 2]$$

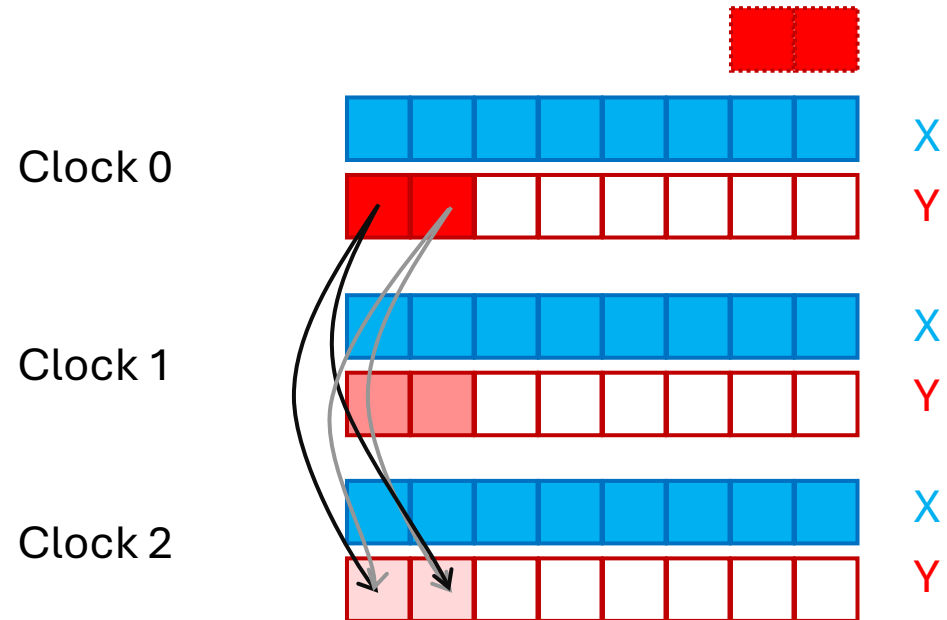


\* Some latency ignored

Signal Processing, DSP  
Notes, and Biquad  
Derivation  
DocDB: PUEO-doc-252

$$y_0[b] = P^{M-1}U_{M-1}(\cos\theta)y_1[b-1] - P^M U_{M-2}(\cos\theta)y_0[b-1] + u_0[b] + \sum_{i=1}^{M-2} P^i U_i(\cos\theta)u_{M-i}[b-1]$$

$$y_1[b] = P^M U_M(\cos\theta)y_1[b-1] - P^{M+1}U_{M-1}(\cos\theta)y_0[b-1] + u_1[b] + P U_1(\cos\theta)u_0[b] + \sum_{i=2}^{M-1} P^i U_i(\cos\theta)u_{M-i+1}[b-1]$$

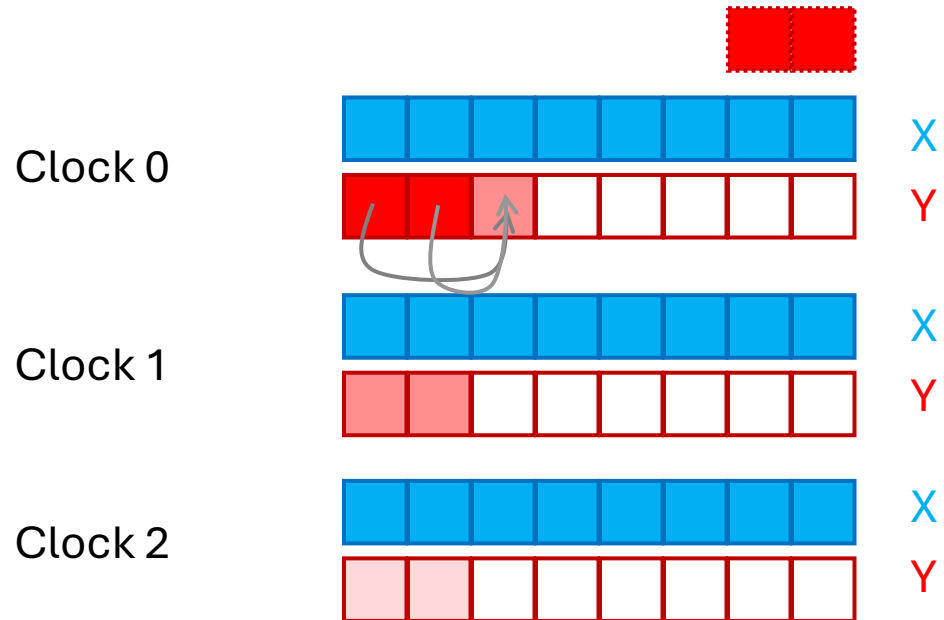


\* Some latency ignored

Signal Processing, DSP  
 Notes, and Biquad  
 Derivation  
 DocDB: PUEO-doc-252

$$H(z) = \frac{b_0 + b_1z^{-1} + b_2z^{-2}}{1 + a_1z^{-1} + a_2z^{-2}}$$

$$y[n] = b_0x[n] + b_1x[n - 1] + b_2x[n - 2] + a_1y[n - 1] + a_2y[n - 2]$$



Now that you are free from dependence outside that cycle of 8, latency is no longer an issue. Just keep doing this for the remaining 6 samples.

\* Some latency ignored



Easy, right?



**Clustered Look-ahead**

Difference Function:  $y[k] = \alpha \cos(\omega) y[k-1] - \beta^2 y[k-2] + x[k]$   
 Transfer Function:  $H_p(z) = \frac{1}{1 - \alpha \cos(\omega) z^{-1} + \beta^2 z^{-2}}$

FROM Feinberg, we can "look back" by using the output from  $k$  and  $k-1$  samples ago, rather than 1 and 2 samples ago in the following way, where  $U_n(\omega)$  are Chebyshev polynomials of the second kind.

(1)  $y[k] = U_M(\cos \theta) P^M y[k-M] - U_{M-1}(\cos \theta) P^{M-1} y[k-(M-1)] + \sum_{i=0}^{M-1} U_i(\cos \theta) P^i x[k-i]$

Next, we work in notation more clearly showing the parallelized nature of this  $M$  and  $N$  and  $M$  is the clock cycle, and  $n$  is the sample in that cycle.  
 Input:  $u_n[b]$   
 Output:  $y_n[b]$

For practical reasons, namely, avoiding putting future incremental computation in the FIR path, we will calculate  $y_0[b]$  and  $y_1[b]$  using  $y[b]$  and  $y[b-1]$ . We expand (1) above for these two cases,  $b' \leq b$ :

$y_0[b] = U_{M-1}(\cos \theta) P^{M-1} y_1[b-1] - U_{M-2}(\cos \theta) P^{M-2} y_0[b-1] + u_0[b] + \sum_{i=1}^{M-2} U_i(\cos \theta) P^i u_{i+1}[b-1]$   
 $y_1[b] = U_M(\cos \theta) P^M y_1[b-1] - U_{M-1}(\cos \theta) P^{M-1} y_0[b-1] + u_1[b] + U_0(\cos \theta) P^0 u_0[b] + \sum_{i=2}^{M-1} U_i(\cos \theta) P^i u_{i+1}[b-1]$

The input terms in the above equations are actually each a decimated (by  $B$ ) FIR, and will now be called  $f[b]$  and  $g[b]$ .

$f[b] = u_0[b] + \sum_{i=1}^{M-2} U_i(\cos \theta) P^i u_{i+1}[b-1]$   
 $g[b] = u_1[b] + U_0(\cos \theta) P^0 u_0[b] + \sum_{i=2}^{M-1} U_i(\cos \theta) P^i u_{i+1}[b-1]$

Note that both of these look back to  $u_i[b-1]$ , and just have different indices for the Chebyshev polynomials. These are FIRs with coefficients  $s = [s_0, s_1, \dots, s_{M-1}]$ ;  $X_i = U_i(\cos \theta) P^i$

As a matrix

$$\begin{bmatrix} y_0[b] \\ y_1[b] \end{bmatrix} = \begin{bmatrix} U_{M-1}(\cos \theta) P^{M-1} & -U_{M-2}(\cos \theta) P^{M-2} \\ -U_{M-1}(\cos \theta) P^{M-1} & U_M(\cos \theta) P^M \end{bmatrix} \begin{bmatrix} y_0[b-1] \\ y_1[b-1] \end{bmatrix} + \begin{bmatrix} f[b] \\ g[b] \end{bmatrix}$$

However, this is more than what two cross coupled DSPs can do at 375 MHz. This means  $y_0[b]$  and  $y_1[b-1]$  will not be ready for use in the  $y_1[b]$  clear cycle yet!!!!

We could look back  $2M$  samples the same way as before, but that would add  $M$  new coefficients to the input calculation from the Chebyshev polynomial. Instead, we can recuse the equation we have, expanding out  $y_1[b-1]$ .

$$\begin{bmatrix} y_0[b] \\ y_1[b] \end{bmatrix} = \begin{bmatrix} \alpha & \beta \\ -\delta & \delta \end{bmatrix} \begin{bmatrix} -\alpha & \beta \\ -\delta & \delta \end{bmatrix} \begin{bmatrix} y_0[b-2] \\ y_1[b-2] \end{bmatrix} + \begin{bmatrix} f[b-1] \\ g[b-1] \end{bmatrix} + \begin{bmatrix} f[b] \\ g[b] \end{bmatrix}$$

$$\begin{bmatrix} y_0[b] \\ y_1[b] \end{bmatrix} = \begin{bmatrix} -(\beta\gamma - \alpha^2) & \beta(\delta - \alpha) \\ -\gamma(\delta - \alpha) & (\delta^2 - \beta\delta) \end{bmatrix} \begin{bmatrix} y_0[b-2] \\ y_1[b-2] \end{bmatrix} + \begin{bmatrix} -\alpha f[b-1] + \beta g[b-1] + f[b] \\ -\delta f[b-1] + \delta g[b-1] + g[b] \end{bmatrix}$$

$$\begin{bmatrix} y_0[b] \\ y_1[b] \end{bmatrix} = \begin{bmatrix} C_0 & C_1 \\ C_2 & C_3 \end{bmatrix} \begin{bmatrix} y_0[b-2] \\ y_1[b-2] \end{bmatrix} + \begin{bmatrix} F[b] \\ G[b] \end{bmatrix}$$

$F[b] = D_{FF} f[b-1] + D_{FG} g[b-1] + f[b]$   
 $G[b] = E_{GF} f[b-1] + E_{GG} g[b-1] + g[b]$

Coefficients	$D_{FF} = -\alpha$	$\alpha = -U_{M-2}(\cos \theta) P^M$
$X_i = U_i(\cos \theta) P^i$ ; $i=0 \dots M-1$	$D_{FG} = \beta$	$\beta = U_{M-1}(\cos \theta) P^{M-1}$
$D_{FF}, D_{FG}, E_{GF}, E_{GG}$	$E_{GF} = -\delta$	Where $\delta = -U_{M-1}(\cos \theta) P^{M-1}$
$C_0, C_1, C_2, C_3$	$E_{GG} = \delta$	$\delta = U_n(\cos \theta) P^M$
$b'_1, b'_2$	$C_0 = \alpha^2 - \beta\gamma$	
$a'_1, a'_2$	$C_1 = \beta(\delta - \alpha)$	
$k$	$C_2 = \gamma(\delta - \alpha)$	
	$C_3 = \delta^2 - \beta\delta$	

Note that all coefficients above are determined by  $P_2, P_3, P_4, k$  OR  $a'_1, a'_2, b'_1, b'_2, k$

DO NOT ERASE

Signal Processing,  
 DSP Notes, and  
 Biquad Derivation  
 DocDB: PUEO-  
 doc-252

# Well, no.

- PUEO's fabrication is currently underway!
- Expecting to integrate the instrument here in Chicago early next year
- Plan to launch in December 2025
- Significant improvements to measurement of diffuse UHE neutrino flux
- Well prepared to measure transient sources

**PUEO will probe the some of the highest energy phenomenon in the universe!**







# Backup

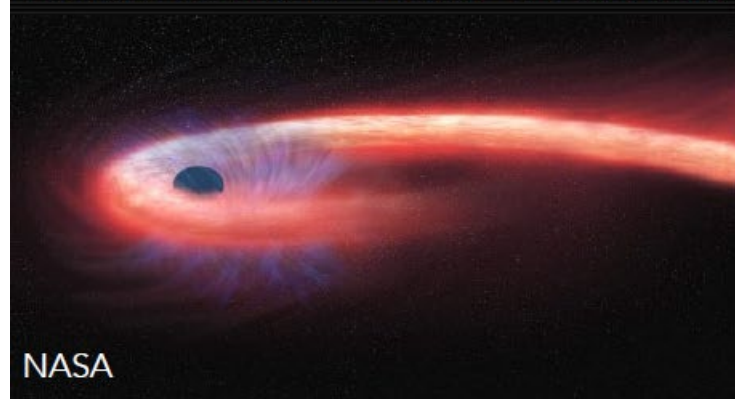
## NEUTRON STAR MERGERS?



## ACTIVE GALACTIC NUCLEI?



## TIDAL DISRUPTION EVENTS?



## GAMMA RAY BURSTS?

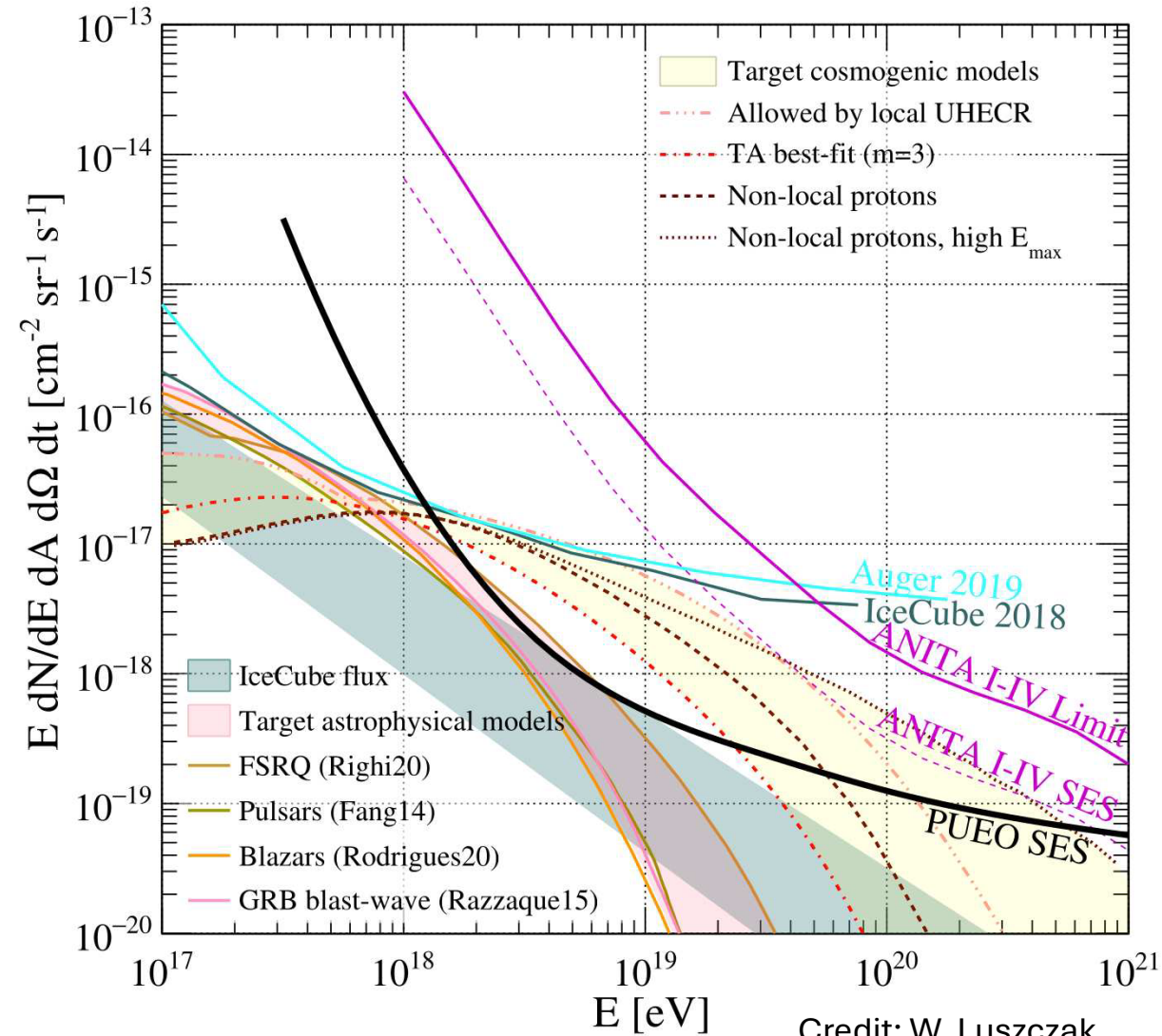


Adapted from K. Hughes

- Sensitive to the lower frequency (50-500 MHz) components present in air showers
  - Overlaps frequency range with main instrument, providing independent measurement
- Enhanced measurement of tau-neutrino induced EASs
- Measurements of cosmic ray induced EASs
  - Understanding neutrino signal background
  - Opportunity to better characterize events like the ANITA “mystery events”

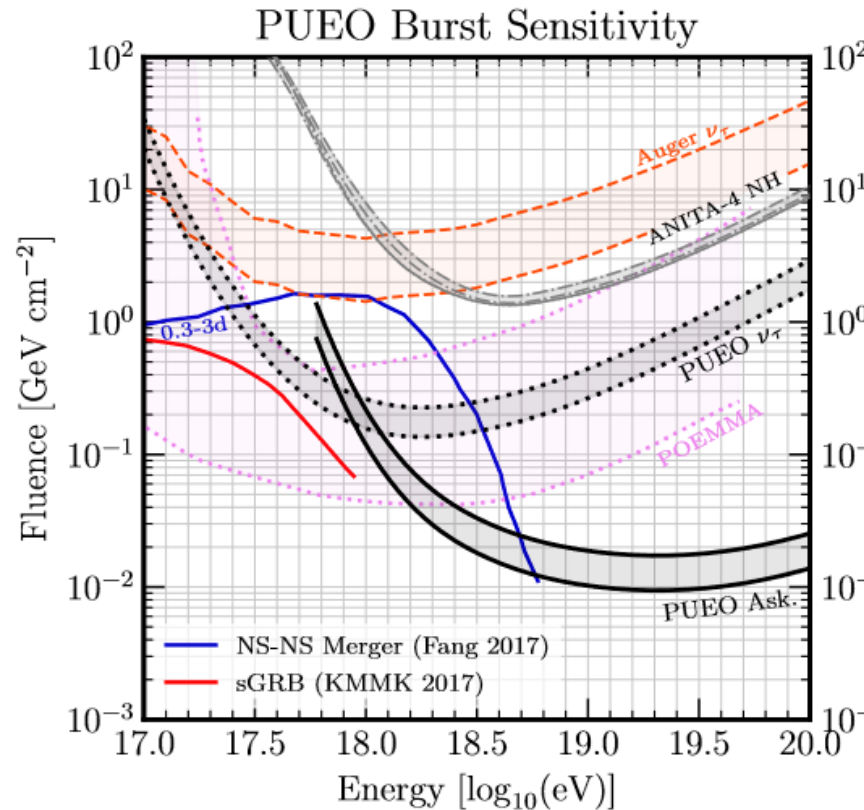


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  - Exclude or measure a number of cosmogenic models
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  - Astrophysical production models

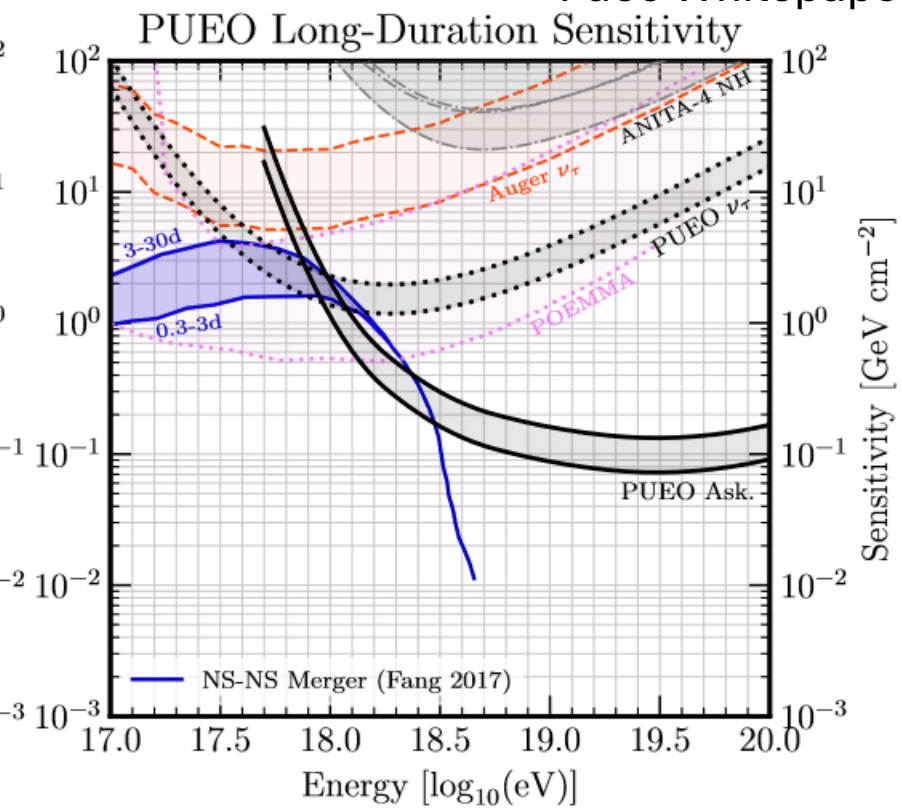


Credit: W. Luszczyk, ICRC 2023

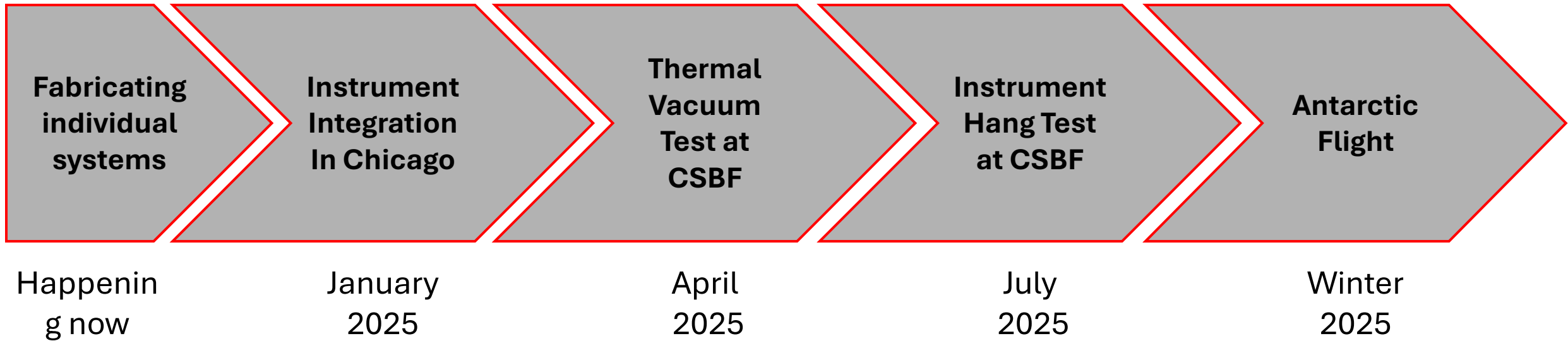
PUEO's large instantaneous aperture makes it ideal for transient searches within its field-of view!

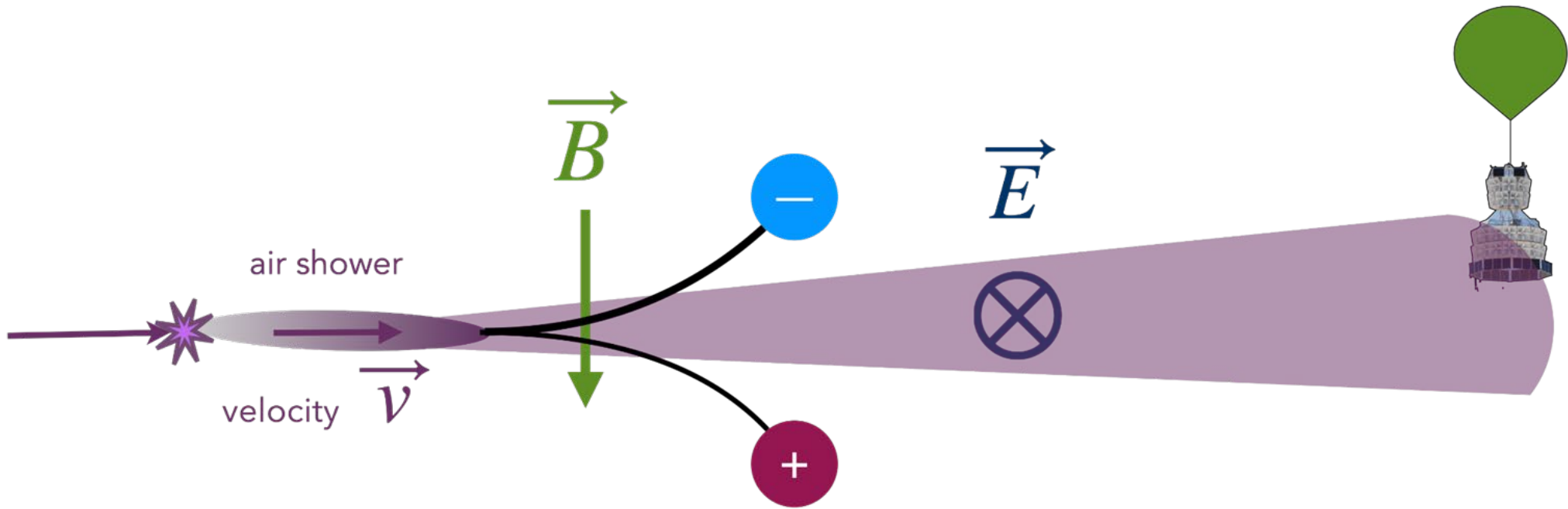


Burst sensitivity is for transients lasting a few hours.



Long duration sensitivity is for transients lasting around the same length as the flight, so effective area is averaged over the full flight.





Credit: Stephanie Wissel